Chapter 3: Characterization of Communication Signals and Systems





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Spectral Characteristics of Digitally Modulated Signals

- Generally, the available channel bandwidth is *limited*
- In the selection of the modulation methods, spectral content of digitally modulated signals be determined
 - This helps to take the effect of BW constraint into account
- A digitally modulated signal is a stochastic process since the information sequences are random
- Need to determined power spectral density (PSD) of these processes
- From PSD one can find the channel bandwidth required to transmit the information-bearing signals



• Consider a linearly modulated band-pass signal given by

$$s(t) = Re\left[v(t)e^{j2\pi f_c t}\right]$$

- Where v(t) is the equivalent low-pass signal
- Autocorrelation function of s(t) is

$$\phi_{\rm ss}(\tau) = Re\left[\phi_{\rm vv}(\tau)e^{j2\pi f_c\tau}\right]$$

- And its Fourier transform yields the desired expression for the PSD $\Phi_{\rm ss}({\rm f})$ as

$$\Phi_{ss}(f) = \frac{1}{2} \left[\Phi_{vv}(f - f_c) + \Phi_{vv}(-f - f_c) \right]$$

• Where $\Phi_{vv}(f)$ is the PSD of v(t)



 To obtain the spectral characteristics of the bandpass signal s(t), it suffices to determine the autocorrelation function and power spectral density of the equivalent lowpass signal v(t)



• Consider a *linear digital modulation* method for which v(t) is represented in the general form

$$v(t) = \sum_{n=-\infty}^{\infty} I_n g(t - nT)$$

- Where:
 - {*I_n*} represents the sequence of symbols resulting from mapping k-bit blocks into corresponding points
 - 1/T = R/k symbols/s is the transmission rate
- The autocorrelation function of v(t) is

$$\phi_{vv}(t+\tau,t) = \frac{1}{2} E \Big[v^*(t) v(t+\tau) \Big]$$

= $\frac{1}{2} \sum_{m=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} E \Big[I_n^* I_m \Big] g^*(t-nT) g(t+\tau-mT)$



 Assuming the sequence of information symbols {*I_n*} is widesense stationary with mean µ_i and autocorrelation function

$$\phi_{ii}(m) = \frac{1}{2} E \Big[I_n^* I_{n+m} \Big]$$

• Then the autocorrelation of v(t) will be

$$\phi_{vv}(t+\tau,t) = \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \phi_{ii}(m-n)g^*(t-nT)g(t+\tau-mT)$$
$$= \sum_{m=-\infty}^{\infty} \phi_{ii}(m)\sum_{n=-\infty}^{\infty} g^*(t-nT)g(t+\tau-nT-mT)$$

• The 2nd summation is periodic with T



• Thus the autocorrelation function is also periodic, i.e.,

$$\phi_{vv}(t+\tau+T,t+T) = \phi_{vv}(t+\tau;t)$$

• Further the mean of v(t)

 $E[v(t)] = \mu_i \sum g(t - nT)$ is also periodic with period T

- v(t) is a stochastic process where both the mean and autocorrelation function are periodic
 - Called a cyclostationary or a periodically stationary process
- To compute the PSD of $\phi_{\nu\nu}(t+\tau,t)$, we eliminate the dependence on variable t by averaging over a single period



$$\begin{split} \bar{\phi}_{vv}(\tau) &= \frac{1}{T} \int_{-T/2}^{T/2} \phi_{vv}(t+\tau;t) dt \\ &= \sum_{m=-\infty}^{\infty} \phi_{ii}(m) \sum_{n=-\infty}^{\infty} \frac{1}{T} \int_{-T/2}^{T/2} g^*(t-nT) g(t+\tau-nT-mT) dt \\ &= \sum_{m=-\infty}^{\infty} \phi_{ii}(m) \sum_{n=-\infty}^{\infty} \frac{1}{T} \int_{-T/2-nT}^{T/2+nT} g^*(t-nT) g(t+\tau-mT) dt \end{split}$$

• In the above expression, the integral can be interpreted as the time autocorrelation of the function *g*(*t*)

$$\phi_{gg}(\tau) = \int_{-\infty}^{\infty} g^*(t)g(t+\tau)dt$$

• So that

$$\bar{\phi}_{vv}(\tau) = \frac{1}{T} \sum_{m=-\infty}^{\infty} \phi_{ii}(m) \phi_{gg}(\tau - mT)$$



• The average PSD of v(t) is the Fourier transform of the average of its autocorrelation which may be expressed as

$$\Phi_{vv}(f) = \frac{1}{T} \left| G(f) \right|^2 \Phi_{ii}(f)$$

• Where G(f) is the Fourier transform of g(t) and

$$\Phi_{ii}(f) = \sum_{m=-\infty}^{\infty} \phi_{ii}(m) e^{-j2\pi fmT}$$



- Note that the power spectral density depends on the spectral characteristic of the pulse g(t) and the information sequence {I_n}
- Or, the spectral characteristics of v(t) can be controlled by the choice of g(t) and the correlation characteristics of the information sequence



 Note also that Φ_{ii}(f) is related to the autocorrelation φ_{ii}(m) in the form of an *exponential Fourier series* with φ_{ii}(m) as the Fourier coefficient such that

$$\phi_{ii}(m) = T \int_{-1/2T}^{1/2T} \Phi_{ii}(f) e^{j2\pi fmT} df$$

• Consider the information symbols in the sequence are real and mutually uncorrelated such that

$$\phi_{ii}(m) = \begin{cases} \sigma_i^2 + \mu_i^2 & (m = 0) \\ \mu_i^2 & (m \neq 0) \end{cases}$$



• Where σ_i^2 is the variance of the information sequence; then

$$\Phi_{ii}(f) = \sigma_i^2 + \mu_i^2 \sum_{m=-\infty}^{\infty} e^{-j2\pi fmT}$$

- Which is periodic with period 1/T
- The above can be viewed as the exponential Fourier series of a periodic train of impulses each with an area of 1/T

• I.e,

$$\Phi_{ii}(f) = \sigma_i^2 + \frac{\mu_i^2}{T} \sum_{m=-\infty}^{\infty} \delta(f - \frac{m}{T})$$



• And substituting this in the expression for $\Phi_{\nu\nu}(f)$

$$\Phi_{vv}(f) = \frac{\sigma_i^2}{T} \left| G(f) \right|^2 + \frac{\mu_i^2}{T^2} \sum \left| G\left(\frac{m}{T}\right) \right|^2 \delta\left(f - \frac{m}{T}\right)$$

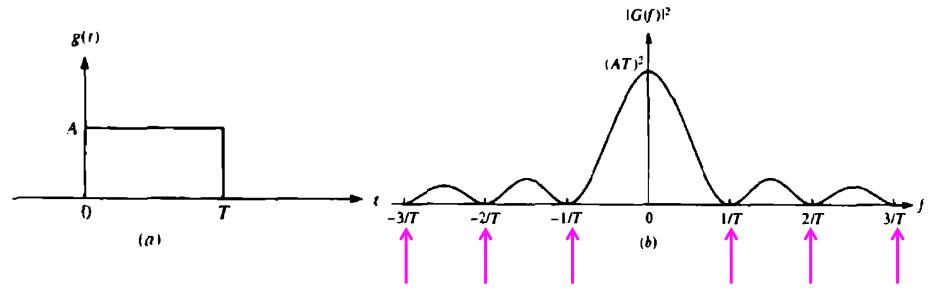
- The first term is a continuous spectrum and its shape depends on the spectral characteristics of signal pulse *g(t)*
- The second expression contains discrete frequency components spaced 1/T apart in frequency
 - Each spectral line has power proportional to |G(f)|² evaluated at f=m/T



- If the information sequences has zero mean, i.e. μ_i =0, the discrete frequency components will vanish
- This property is most desirable for digital modulation and can be achieved when the information sequences are *equally likely* & *symmetrically positioned* in a complex plane



 Example 1: Consider g(t) to be a rectangular pulse as shown in the figure below with Fourier transform |G(f)



Rectangular pulse and its energy density spectrum

$$G(f) = AT \frac{\sin \pi fT}{\pi fT}$$
 and $|G(f)|^2 = (AT)^2 \left(\frac{\sin \pi fT}{\pi fT}\right)^2$



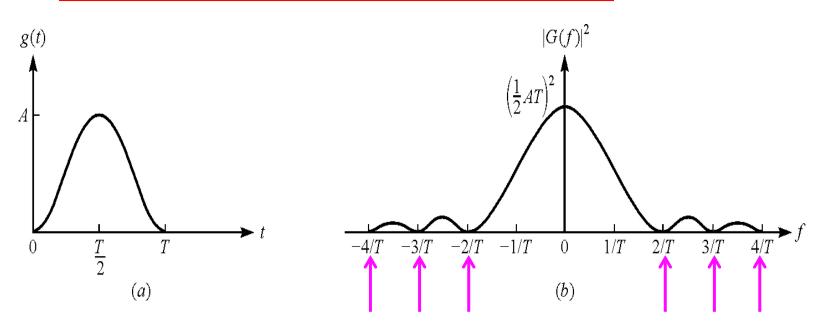
- It contains zeros at multiples of 1/T in frequency and it also decays inversely as the square of the frequency variable
- As a result, all but one of the discrete spectral components in Φ_{νν}(f) vanishes
- Thus, upon substitution for |G(f) from above, we get

$$\Phi_{vv}(f) = \sigma_i^2 A^2 T \left(\frac{\sin \pi f T}{\pi f T}\right)^2 + A^2 \mu_i^2 \delta(f)$$



Example 2: Consider the case where g(t) is a raised cosine pulse

$$g(t) = \frac{A}{2} \left[1 + \cos \frac{2\pi}{T} \left(t - \frac{T}{2} \right) \right] \qquad \text{for} \quad 0 \le t \le T$$



Raised cosine pulse and its energy density spectrum

• Its Fourier transform is given as

$$G(f) = \frac{AT}{2} \frac{\sin \pi fT}{\pi fT(1 - f^2T^2)} e^{-j\pi fT}$$

- Note the spectrum has zeros at f = n/T; $n = \pm 2, \pm 3, \pm 4...$
- Hence, all the spectral components, except those at zero and $f = \pm 1/T$ vanish
- Compared to that of the rectangular pulse, the spectrum has a broader main lobe but the tails decay inversely as *f*⁶



- Spectrum can also be shaped by operations performed on the input information sequence
- Example 3: Consider a binary sequence {b_n} from which we form the symbol

$$I_n = b_n + b_{n-1}$$

- Where the {*b_n*} are assumed to be uncorrelated random variables, each having zero mean and unit variance
- The autocorrelation of the sequence $\{I_n\}$ is

$$\phi_{ii}(m) = E(I_n I_{n-m})$$

$$= \begin{cases} 2 & (m=0) \\ 1 & (m=\pm 1) \\ 0 & (Otherwise) \end{cases}$$



• The PSD of the input sequence is

$$\Phi_{ii}(f) = 2(1 + \cos 2\pi fT) = 4\cos^2\pi fT$$

• Then, the corresponding power spectral density of the (lowpass equivalent) modulated signal becomes

$$\Phi_{vv}(f) = \frac{4}{T} \left| G(f) \right|^2 \cos^2 \pi f T$$

